

amateur radio

Vol. 39, No. 4

APRIL, 1971

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amateur radio

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Reg. Office: 478 Victoria Parade, East Melbourne, Vic., 3002.

Editor:

K. E. PINCOTT VK3AFJ

Publications Committee:

R. Dorin	VK3ZU
Ken Gillespie	VK3GK
Harold Hepburn (Secretary)	VK3APO
Peter Ramsay	VK3ZWN
W. E. J. Roper	VK3ARZ

Circulation—

Jack Kelly VK3AFD

Draughtsmen:

Clem Allan	VK3ZIV
John Blanch	VK3ZOL
John Whitehead	VK3YAC

Enquiries:

Mrs. BELLAIRS, Phone 41-3535, 478 Victoria Parade, East Melbourne, Vic., 3002. Hours: 10 a.m. to 3 p.m. only.

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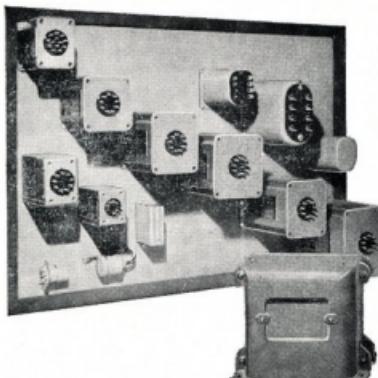
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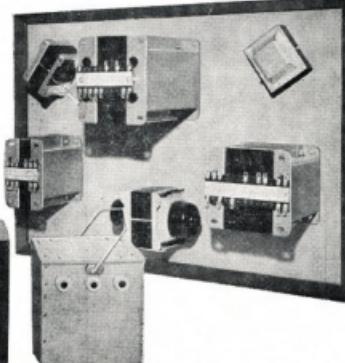
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LM 51

FEDERAL COMMENT

In the January issue I wrote about the Federal Executive's problems revolving round the near impossible situation facing the many honorary officers administering the organisation and this magazine. I gave a brief outline of the facts which brought about the decision to employ a Secretary/Manager. In that same issue there appeared an advertisement for filling this post.

I am very pleased to tell you that the post has now been filled following upon interviews with candidates on a short list selected from all the applications which were received. The successful candidate happened to be in Australia at the time when the post was advertised and it is our good fortune that his services are now available to us.

He is 53-year-old PETER B. DODD, VK6/5/3/1/2C1F, better known perhaps as a past DXer with such call signs as VQ4PBD, VQ5PBD, VQ1PBD, 5H3PBD, 7Q7PBD, G3PBD and many others dating back to 1946 and to pre-war as a listener. He has also operated for a short time as ZL1BDC portable/mobile s.s.b. from a motor caravan in which he and his family travelled overland from Europe. On this safari he operated 4U1ITU and held calls as OE1ZBW and YA1PBD.

In addition to being reasonably well known on the DX bands, he is a Life Vice-President of the Radio Society of East Africa. He served on the Council of that Society, organised Amateur Radio familiarisation exercises for the benefit of local Ministers of at least two African governments, was closely involved with the establishment and progress of the East African Emergency Network allied with communications for the world-renowned annual East African Safari and, when not resident in Nairobi, the Society's headquarters, reminded them that there were such

people as country members. I gather from another source that he was awarded a medal by the Belgian Government for work done during the Congo crisis.

On the general administrating side, Peter Dodd had come up through the ranks of Customs and Excise in East Africa, culminating as Head of the Department in Malawi where he was responsible for establishing it in that country. For a period he was a Director of an Amateur equipment manufacturing company in the U.K. The Selection Committee were satisfied that he would bring to the position almost unique experience with impartial detachment, a wealth of administrative ability and a fund of enthusiasm. We wish him well.

It is fortunate too that we will possess someone capable of effecting a smooth transition from the existing to the new Constitution of the W.I.A. which is mentioned in my Report, to be published in "A.R.", to be considered at the Federal Convention in Brisbane this month. No doubt your Federal Councillor will have informed you about the various motions which are to be debated at this Convention.

However, it is thought that the Convention will give more time in considering the precise plans which will be necessary to effect the change-over to the new W.I.A. Constitution, the I.A.R.U. Region 3 Conference in Tokyo and the I.T.U. World Administrative Radio Conference in Geneva later in the year. I ask you to read these references with care and to observe the work being done on behalf of all Amateurs in this part of the globe.

Once again I seek your support by continuing your interest and by each one of you recruiting at least one more member this year.

MICHAEL OWEN, VK3KI,
Federal President, W.I.A.

A Transistorised Carphone

PART TWO—TRANSMITTER

By L. B. JENKINS,[†] VK3ZBJ, and H. L. HEPBURN,[‡] VK3AFQ

The authors continue this second part of the article with a description of the transmitter and associated circuits. From correspondence received, it appears that boards, diagrams and/or kits of this Carphone will be in demand. Accordingly, work is proceeding along these lines.

The transmitter can conveniently be dealt with in three parts—the exciter/modulator, the driver stage, and the final power amplifier. This grouping is chosen since each module represents a physically separate entity with each module on a separate circuit board. The exciter/modulator is on a p.c.b. 7" x 2", while both driver and p.a. stages are each on boards 3½" x 2".

It must be stated right at the outset that the transmitter is a frequency modulated device and not (as are the popular Vintens, A.W.A.'s and T.C.A.'s) a phase modulated system. The decision to use f.m. rather than p.m. was based mainly on circuit simplicity and ease of adjustment.

Briefly, the main difference between the two methods is that with f.m. the amount of deviation is proportional only to the level of audio drive and is independent of the modulating frequency. With p.m. on the other hand, the amount of deviation is proportional

not only to the audio level, but also the frequency of the modulating signal, the higher the modulating frequency the greater the amount of deviation. Thus in a p.m. system it is necessary in the transmitter to reduce the audio drive as the modulating frequency increases (de-emphasis) and in the associated receiver to increase the audio "highs" to compensate (pre-emphasis). The amounts of pre-emphasis and de-emphasis used in surplus commercial units varies from make to make.

In order that the transmitter now described be compatible with the wide variety of transceivers in Amateur use, it has been necessary to provide some audio shaping in the modulator. Since this shaping has been done with only two fixed resistors and two fixed capacitors and since these can be altered to taste, obtaining audio compatibility presents no problem.

THE EXCITER/MODULATOR

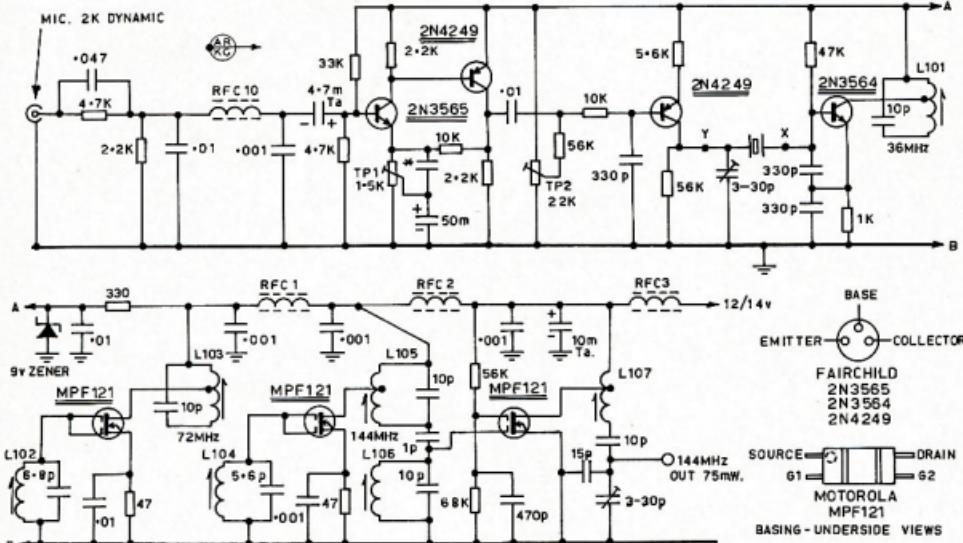
Fig. 6 gives the circuit diagram of the complete exciter. It uses three bipolar

transistors in the audio section, a bipolar oscillator and three protected dual gate MOSFETs as multipliers. These latter devices are quite new on the Australian scene and can best be described as higher dissipation, epoxy packaged 3N140s, but without the problems of static destruction associated with the latter device. They can be handled and soldered into place with no more care than the normal run of transistors.

The MPF121s have been used in this design on two grounds. Firstly, they are the same price as the type of bipolar transistor used to date in multiplier service (2N3564, 2N3565, BF115, etc.), but more importantly they are to be preferred in view of the almost perfect waveform purity that can be obtained. Those who have had experience with bipolars as frequency multipliers will be aware of the difficulties of obtaining a sub harmonic and spurious-free waveform from them.

Input from a medium impedance dynamic or rocking armature micro-

[†] 54 Tennyson Street, Highett, Vic., 3190.
[‡] 4 Elizabeth Street, East Brighton, Vic., 3187.



* May be needed to adjust frequency response.

phone (a Zephyr 2SE 2,000 ohm p.t.t. is standard in both authors' equipment and is thoroughly recommended) is shaped by the 0.047 μ F/4.7K ohms and 0.1 μ F/2.2K ohms combinations. If any other microphone is used, or if a different audio characteristic is required then some adjustment to these values will be needed. RFC10, which consists of a single wire through a Neosid F29 tuning slug, and the associated 1,000 μ F capacitor decouple the base of the 2N3565 for r.f.

The 2N3565/2N4249 bipolar combination provides ample audio gain, this gain being adjustable through TP1 (1.5K) which acts as a deviation control. Audio is applied to the base of the 2N4249 modulator bipolar whose base d.c. voltage is adjustable by means of TP2 (22K). This variable resistor allows control over both frequency and speech linearity.

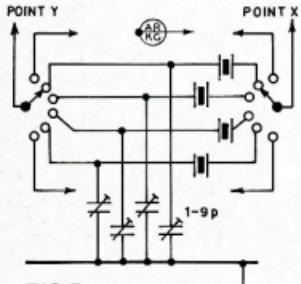


FIG.7.
CRYSTAL SWITCHING

The 2N4249 modulator acts in effect as a variable capacitor which is in series with the crystal and ground. Any variation of the voltage on the base of the modulator (no matter whether it be d.c. or audio) varies the capacity in series with the crystal which, in turn, varies the frequency of oscillation. For accurate setting of centre frequency and reliable operation it is essential that the crystal have a low equivalent series resistance.

It may be worth noting at this point that if the f.m. modulator bipolar is omitted and the crystal grounded through, say, 30 pF, then the r.f. generating side of things can be used to drive an a.m. final while the 2N3565/2N4249 combination can be used as a microphone pre-amplifier.

The oscillator uses crystals in the 12 MHz range, the exact frequency being obtained by dividing the required output frequency by twelve.

Fig. 7 gives the circuitry used for multi-channel operation, each crystal having its own trimmer for frequency adjustment. The trimmers recommended are the 1 to 9 pF Shilnne types sold through the VK3 W.I.A. components facility. These trimmers are also used in the driver stage of the transmitter.

Output from the oscillator at 36 MHz. is transferred by means of the mutually coupled pair L101/L102 to the parallel gates of the first MPF121 doubler. Output from this stage on 72 MHz. goes through the second mutually

coupled pair L103/L104 to the parallel gates of the second MPF121 doubler. Again a pair of coils is used to transfer the 144 MHz. output to the third MPF121. Some capacitive top coupling is used in this case. The third MPF121 is used as an amplifier and has about 7 volts applied to its second gate. A series tuned circuit in the drain uses a capacitive divider to give a 50 ohm output impedance. The trimmer at the bottom of this divider is a standard Philips 3-30 pF unit.

Setting up of the Unit

This is simple but does require some form of output indicator, a milliammeter, and an absorption wavemeter/g.d.o. covering 30 to 80 MHz. A circuit of a suitable output indicator is given in Fig. 10. It consists of a 47 ohm load resistor, a germanium diode such as an OA91 and a voltmeter. Assuming an output of 100 mW. from the exciter, the rectified d.c. will be about 24 volts. If the indicator is used to set up the driver and p.a. stages then voltages of respectively 7-8 and 20-25 will be encountered.

A carbon resistor must be used and not a wire wound one. A one watt resistor is suitable for the exciter (and even possibly for the driver), but the best overall solution is to parallel ten 470 ohm one watt resistors to give a power handling capacity of ten watts. The indicator can then be used for the p.a. as well. Keep all connections as short as possible.

Bear in mind that the above indicator is just that. If a proper measuring power meter is required then a kit of parts for a fully shielded, two range (0.5 and 0-50 watts) power meter put out by Horwood Electronics in Melbourne is recommended. They can also be purchased fully made up and tested from Radio Parts Pty. Ltd. in Melbourne.

The commissioning procedure is as follows. Set the deviation control (TP1) to minimum, i.e. with the slider earthed. Put dummy load across the output of the exciter. This load may consist simply of a 47 ohm resistor, or

the indicator described or a proper 50 ohm power meter. Apply 12 volts through a 0-250 mA. meter. Set TP2 so that the voltage between the collector of the 2N4249 modulator transistor and earth is about 5 volts. TP2 should be about the middle of its range. At this stage the current drawn should be around 20 mA. and the oscillator may or may not be going.

Couple an absorption wavemeter (or g.d.o. in the wavemeter position) to the oscillator collector coil L101 and adjust its core until output on 36 MHz. is obtained. Then set the wavemeter to 72 MHz., couple it to L103 and adjust the cores of L102 and L103 for maximum output. Note that as each of the cores is adjusted, and as output comes up, the total current drawn will increase, each of the MPF121 stages pulling some 20-25 mA. as it comes on to resonance.

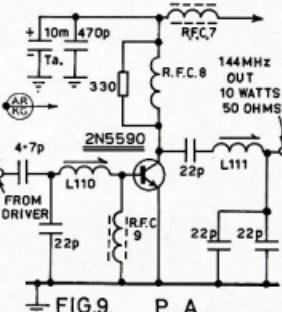


FIG.9. P.A.

Now set the 3-30 pF. output trimmer to about 1" mesh and adjust the cores of L105, L106, L107 and the output trimmer until some indication of output from the exciter is seen. At this stage go back over all the coils and adjust their cores for maximum indicated output. When on tune, at least 75 mW. at 144 MHz. should be available.

Using a receiver on the appropriate channel as a monitor, the modulator may now be adjusted. Set TP1 to full open (slider at the emitter end), connect the appropriate microphone and, while speaking into the microphone, adjust TP2 for the most intelligible speech in the monitor. Frequency can then be set using the crystal trimmer by zero beating against a station known to be on the correct frequency. The unit may then be put to air for final adjustments to TP1 for deviation, TP2 for speech linearity and the crystal trimmer for frequency, bearing in mind that the last two adjustments interact.

THE DRIVER STAGE

The driver stage uses a Motorola 2N5589 (MIL1601) to raise the power level to 1-1½ watts. Fig. 8 gives the appropriate circuit diagram.

Input from the exciter at 50 ohms is matched to the transistor base by the two 1-9 pF trimmers and L108, while the output impedance is brought up to 50 ohms by means of L109 and its associated capacitors.

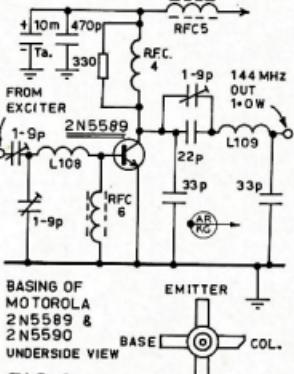


FIG.8.
DRIVER

A low value resistor is used across the collector choke to reduce Q and inhibit parasitic oscillation. The h.t. power is decoupled by means of RFC5 and the 470 pF/10 μ F combination.

RFC6 at the base of the transistor consists of a single wire running through a half-inch length of ferrite rod which has high losses at the frequency of operation. Use of high frequency material such as the Neosid F29 slugs used as decoupling devices elsewhere in the design is to be avoided. In the absence of suitable ferrite, a 1 watt 47 ohm resistor can be substituted with only a small drop in overall efficiency.

The 470 pF, h.t. decoupling capacitor is a normal disc ceramic and the 10 μ F, a tantalum, but all other capacitors in the signal circuits are Philips ceramic beads. The trimmers are the Shimmei type previously mentioned.

Setting up is relatively simple. A 50 ohm dummy load is connected to the output and all variable capacitors set to full capacity. Drive is applied from the exciter together with an initial h.t. of 3-4 volts fed in through a 0.500 mA. meter. The input (series) trimmer is reduced in capacity until the current drain begins to rise and output is indicated. All three capacitors are then adjusted for maximum output. The h.t. is then raised to, say, 9 volts and the trimmers adjusted for maximum output. Finally, full h.t. is applied and again the three trimmers adjusted for maximum output.

Note that at full h.t. the 1-9 pF. trimmer between L108 and earth should be between half band and full capacity, while the series trimmer should be between half and zero capacity. The current drawn by the driver stage alone should be about 250 mA. Currents grossly in excess of this are an indication either of mistuning or of parasitic oscillation. Power output should be at least 1½ watts.

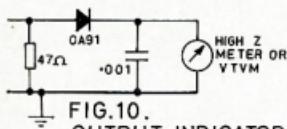


FIG.10.
OUTPUT INDICATOR

THE OUTPUT STAGE

A Motorola 2N5590 (MM1602) is used to raise the output power to the 10 watt level. Note that this is r.f. power output and not d.c. power input.

In most respects the p.a. stage is a copy of the driver stage except that it uses fixed capacities and variable inductance rather than the other way round. All the capacitors except the 470 pF/10 μ F tantalum are Philips beads and two are paralleled at the output to increase the power handling capacity. Note that the normal disc ceramics are not intended to carry large r.f. currents and, if used, will run hot or blow. The "lossy" ferrite RFC technique is again used in the base of the transistor.

Tune up follows the same lines as the driver. The cores of L110 and L111 are set full in, a 50 ohm load connected,

drive is applied and a low level of h.t. fed in through a 0-2 amp. meter. The cores of the two coils are adjusted for maximum output. H.t. is then raised in two or three steps to maximum, at each step the coil slugs being adjusted for maximum output consistent with the lowest collector current drain. If, at any time, the collector current rises at a more rapid rate than the r.f. output is rising, then it is possible that the stage is breaking into oscillation or is being mistuned. As a guide, at 10 watts r.f. output and a 13.5 volt rail, the p.a. should draw no more than 1 amp.

COIL DATA

- L101—20 turns 26 B. & S. enam., tapped 6 turns, close wound on Neosid 722/1 former, F29 slug.
- L102—20 turns 26 B. & S. enam., close wound on Neosid 722/1 former, F29 slug.
- L103—10 turns 23 B. & S. enam., tapped 3 turns, close wound on Neosid 722/1 former, F29 slug.
- L104—10 turns 23 B. & S. enam., close wound on Neosid 722/1 former, F29 slug.
- L105—4½ turns 18 B. & S. tinned copper, spaced $\frac{1}{2}$ ", tapped 2 turns, on Neosid 722/1 former, F29 slug.
- L106—4½ turns 18 B. & S. tinned copper, spaced $\frac{1}{2}$ ", on Neosid 722/1 former, F29 slug.
- L107—5½ turns 18 B. & S. tinned copper, spaced $\frac{1}{2}$ ", tapped $\frac{1}{2}$ turns, on Neosid 722/1 former, F29 slug.
- L108 (driver)—4 turns 18 B. & S. tinned copper, air cored, 5/16" i.d., spaced $\frac{1}{2}$ ".
- L109 (driver)—5 turns 18 B. & S. tinned copper, air cored, 5/16" i.d., spaced $\frac{1}{2}$ ".
- L110 (p.a.)—12 turns 18 B. & S. tinned copper, spaced $\frac{1}{2}$ ", on Neosid 722/1 former, F29 slug.
- L111 (p.a.)—3½ turns 18 B. & S. tinned copper, spaced $\frac{1}{2}$ ", on Neosid 722/1 former, F29 slug.
- RFC1, 2, 3, 5, 7, 10—Single wire through F29 slug.
- RFC4—6 turns 23 B. & S. enam., close wound on $\frac{1}{2}$ " i.d. air cored, $\frac{1}{2}$ " long.
- RFC6—Single wire through ferrite rod $\frac{1}{2}$ " long (or 47 ohm resistor).

RFC8—3 turns 18 B. & S. tinned copper, $\frac{1}{2}$ " i.d., spaced to occupy $\frac{1}{2}$ " length.

RFC9—Single wire through ferrite rod $\frac{1}{2}$ " long (or 33 ohm resistor).

GENERAL

While the designs presented for both the receiver and transmitter are well up with the current state of the art, they are not so far "out" that they are impractical to build because the key components are obtainable. The two key components in this case are the Toyo 10M-2A-1 filter which is marketed in Australia by Arbor Pty. Ltd., of 232 Bell Street, Coburg, Vic., and the AWM1272 and 1306 which can be obtained from A.W.V. in Sydney. The 455 KHz, i.f. transformers used are "Rapar 6" replacement transformers from Radio Parts Pty. Ltd. in Melbourne (who also stock the Fairchild transistors), while the 2N5589/90, and the MC1454 are from Total Electronics of 239 Bay Street, Brighton, Vic., 3186. All other "bits" are normal components held by the VK3 W.I.A. new components service at P.O. Box 65, Mt. Waverley, Vic.

At the end of Part One it was stated that boards, diagrams and/or kits would be made available if required. From subsequent correspondence it appears that such requirement exists and, accordingly, work is proceeding to do this. Further details can be obtained from either of the authors.

In conclusion there are a couple of points that may be of interest. It was stated earlier in this article that the MFPI12s had been used because of their ability to give excellent waveform. The complete transmitter, running 10 watts into a dummy load, when checked with a Philips v.h.f. sampling c.r.o. showed no sign of sub harmonic content and an excellent waveform, indicating minimal higher order harmonics. Secondly, it should be noted that the driver and p.a. transistors are rated for infinite s.w.r., i.e. they should work into an open circuit or a short circuit. Whilst most definitely not recommended as normal operating procedure, such a specification does much to reduce fears of catastrophic failure of relatively expensive devices due to accidental short or open output conditions.

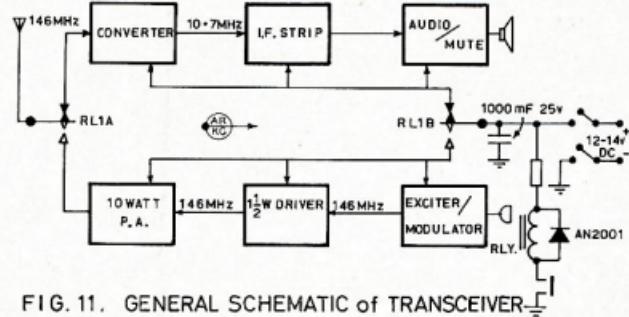


FIG.11. GENERAL SCHEMATIC of TRANSCEIVER

THE DECIBEL AND DECIBELS V. % DISTORTION

LECTURE NO. 11

THE DECIBEL

In communications systems it is convenient when making measurements or calculations to express the RATIO between any two amounts of electric or acoustic power in units on a logarithmic scale.

The DECIBEL (1/10th of the BEL) on the Briggs (Base 10) scale is in almost universal use, although sometimes the NEPER on the Napierian base-e-scale is used.

Because voltage and current are related to power by impedance, both the decibel and the neper can be used to express voltage and current ratios, provided care is taken to account for the impedances associated with them.

In a similar manner, corresponding acoustical powers may be compared.

It must be understood, thoroughly, that both the decibel and the neper are RATIOS and have no meaning unless a reference is stated. For instance, it makes sense if we state that the ratio of one thing to another is 10 to 1, but it is meaningless if we simply state that the ratio is 10, because we no longer have a reference.

In radio work the decibel is used almost exclusively to express ratios and in dealing with Audio Frequency power it is almost universal to use a reference level of 1 milliwatt power in 600 ohms, known as 0 dbm, or zero dbm. In this context 0, or zero, does not mean nothing or nil but the transition between powers less than or greater than 1 milliwatt in 600 ohms (0 dbm).

The number of decibels (Ndb) corresponding to the ratio between two amounts of power P_1 and P_2 is

$$\text{Ndb} = 10 \log_{10} \frac{P_1}{P_2}$$

when two voltages E_1 and E_2 , or two currents I_1 and I_2 , operate in the same or equal impedances,

$$\text{Ndb} = 20 \log_{10} \frac{E_1}{E_2}$$

$$\text{and Ndb} = 20 \log_{10} \frac{I_1}{I_2}$$

If E_1 and E_2 , or I_1 and I_2 , operate in unequal impedances,

Ndb =

$$20 \log_{10} \frac{E_1}{E_2} \pm 10 \log_{10} \frac{Z_1}{Z_2} \pm 10 \log_{10} \frac{K_1}{K_2}$$

and Ndb =

$$20 \log_{10} \frac{I_1}{I_2} + 10 \log_{10} \frac{Z_1}{Z_2} + 10 \log_{10} \frac{K_1}{K_2}$$

where Z_1 and Z_2 are the absolute magnitude of the corresponding impedances and K_1 and K_2 are the values of power factor for the respective impedances.

- Continuing the series of lectures by C. A. Cullinan, VK3AXU, at Broadcast Station 3CS for students studying for a P.M.G. Radio Operator's Certificate.

It will be seen from the above formulae that power, voltage and current ratios may be expressed logarithmically in decibels irrespective of whether the impedances are equal or unequal.

It is possible to convert decibels to nepers and vice-versa.

Multiply decibels by 0.1151 to find nepers.

Multiply nepers by 8.686 to find decibels.

DECIBELS V. % DISTORTION

In its Standards for the Technical Equipment and Operation of Medium Frequency Broadcasting Stations, second edition, 18th June, 1968, the Australian Broadcasting Control Board requires that the harmonic distortion in equipment be expressed in a percentage of the effective value of the fundamental audio frequency voltage and the harmonic voltages present in the output.

However, in recent times there has been a tendency for some authorities and manufacturers of equipment to

express harmonic distortion in decibels instead of in percentage, and until one becomes familiar with this it can be very inconvenient.

Therefore a conversion table has been prepared showing the equivalent distortion for a given db. ratio covering 10% to 0.1% distortion.

The full output voltage is the reference of 0 db. = 100%.

Decibels	Distortion %	Decibels	Distortion %
-20	10.000	-41	0.8913
-21	8.913	-42	0.7943
-22	7.943	-43	0.7079
-23	7.079	-44	0.6310
-24	6.310	-45	0.5623
-25	5.623	-46	0.5012
-26	5.012	-47	0.4467
-27	4.467	-48	0.3981
-28	3.981	-49	0.3548
-29	3.548	-50	0.3162
-30	3.162	-51	0.2818
-31	2.818	-52	0.2512
-32	2.512	-53	0.2239
-33	2.239	-54	0.1995
-34	1.995	-55	0.1778
-35	1.778	-56	0.1585
-36	1.585	-57	0.1413
-37	1.413	-58	0.1259
-38	1.259	-59	0.1222
-39	1.222	-60	0.1000

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A 20W. 576 MHz. VARACTOR MULTIPLIER TRANSMITTER

R. J. HALLIGAN,* VK3AOT/T

After an examination of the theory of varactor frequency multiplication, two practical frequency quadruplers are presented. The first will deliver 10 watts FM/CW at an efficiency of 33%, while the second will deliver 20 watts FM/CW at an efficiency of 50%. Operation with amplitude modulated signals is also possible.

Varactor diodes are a class of semiconductor device intended for power-frequency multiplication at v.h.f. and above. Circuits are characterised by the absence of any d.c. power input, high r.f. output to r.f. input efficiencies, and simple construction. Using varactor techniques powers in excess of 300W. at 100 MHz, and 25W. at 1,000 MHz. have been obtained.

The response of varactor multiplier circuits to amplitude modulated inputs is dependent on the power level, modulation percentage and type of diode. Most designs are capable of providing results acceptable to the Amateur. Some of the more recently developed diodes have been used commercially for the frequency multiplication of television signals, an application requiring a high degree of linearity.

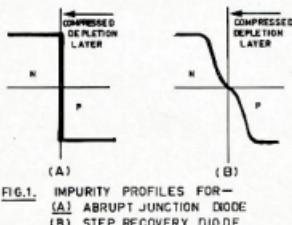


FIG. 1.—COMPARISON OF IMPURITY PROFILES FOR—
(A) ABRUPT JUNCTION DIODE
(B) STEP RECOVERY DIODE

Fig. 1.—Comparison of impurity profiles for abrupt junction and step recovery diodes.

* 41 Windsor Avenue, Mt. Waverley, Vic., 3149.

THEORY OF OPERATION

Abrupt Junction Varactors.—Early varactor diodes relied on the capacitance-voltage non-linearity characteristic of an abrupt P-N junction. Such a junction is the result of a constant resistivity profile in both the P and N regions. See Fig. 1. The dependence of capacitance on voltage is given by equation 1.

$$C_0 = \frac{C_0}{(1 + V/\phi)^{1/2}} \quad \dots \quad (1)$$

where C_0 is the voltage dependent junction capacitance.

C_0 is the capacitance at zero bias.

V is the reverse bias voltage across the varactor.

ϕ is the contact potential, approx. 0.5 for silicon.

In order to ensure high diode 'Q' and therefore good efficiency, series resistance and therefore resistivity must be kept low. However, low resistivity results in low breakdown voltage, giving rise to significant power limitations.

There are also limitations in the response of abrupt junction varactors to amplitude modulated signals. The harmonic generation mechanism as given by equation 1 is voltage dependent, therefore the abrupt junction varactor cannot react to both high and low level signals with the same efficiency. Of greater importance is the variation of varactor capacitance with changes in signal level, leading to circuit detuning during the amplitude modulated cycle. This undesirable mechanism causes the "switching" commonly seen with varactor multipliers. In some cases the varactor will

even act as the active element of a parametric oscillator, with the input signal acting as pump source. When this occurs an unwanted discontinuity or oscillation appears on the amplitude modulated waveform.

Step-Recovery Varactors.—More modern devices are not subject to these power and linearity limitations. These devices are constructed so that the resistivity of the material peaks sharply in the vicinity of the junction (depletion region), but is low elsewhere. A typical impurity profile for this type is also shown in Fig. 1.

The effect of this construction is to reduce the dependence of junction capacitance on voltage so that this is no longer the dominant mechanism for the generation of harmonics. Instead, harmonics are generated by a pulse of reverse current resulting from the return of stored carriers. This is known as the step-recovery effect.

AVAILABLE DIODES

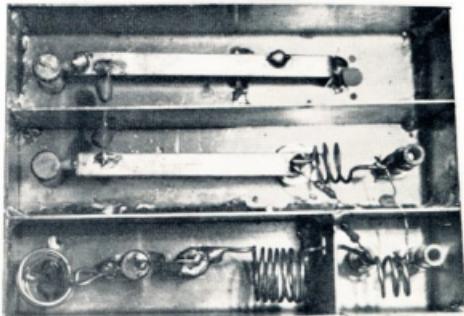
The table lists the characteristics of some varactor diodes which are available. Also listed are some transistors, the collector-base junctions of which can be used for varactor multiplication.

A PRACTICAL 576 MHz. QUADRUPLER

The circuit of a practical quadrupler is shown in Fig. 2. L1-C1 and L2-C2 form a simple double tuned circuit matching network at 144 MHz. Currents at that frequency are caused to flow in D1, which is effectively a capacitor. However, since this capacitance is non-linear, harmonics of 144 MHz. are produced. As is common with



Top view of improved doubler-doubler circuit.



Bottom view of improved doubler-doubler circuit.

harmonic generators, the second harmonic is strongest, with subsequent harmonics progressively diminishing in amplitude. It is quite feasible to simply couple the diode to a tuned circuit at 576 MHz, and extract energy at this frequency. However, because of the small amplitude of the fourth harmonic efficiency would be low.

Efficiency can be improved by the addition of series resonant idler circuits at 288 MHz. (L3-C3) and 432 MHz. (L4-C4). These idlers re-circulate the harmonics, which are mixed with other components or multiplied within the diode, so enhancing 576 MHz. output.

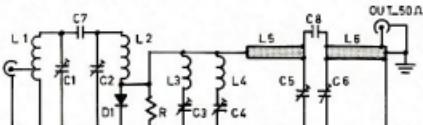


FIG. 2. 144-576 VARACTOR QUADRUPLER

L1-8 turns 18 s.w.g., $\frac{1}{8}$ -inch I.d., tapped $\frac{1}{2}$ turns from cold end, spaced $\frac{1}{8}$ -inch.
 L2-6 turns 18 s.w.g., $\frac{1}{8}$ -inch I.d., spaced $\frac{1}{2}$ -inch.
 L3-3 turns 18 s.w.g., $\frac{5}{16}$ -inch I.d.
 L4-1 turn 18 s.w.g., $\frac{5}{16}$ -inch I.d.
 L5-3 inch x $\frac{3}{16}$ -inch 22 s.w.g. brass strip, $\frac{3}{8}$ -inch above box.
 L6- $\frac{3}{8}$ -inch x $\frac{3}{16}$ -inch 22 s.w.g. brass strip, $\frac{3}{8}$ -inch above box.

Resistor R serves to develop self-bias for the diode. While the varactor is primarily a variable capacitor for harmonic generation, it does conduct at one peak of every cycle. The subsequent d.c. current flow through R establishes a bias point for the diode.

L5-C5 and L6-C6 are resonant at 576 MHz. and attenuate undesired products. The load is tapped onto L6 at a point such as to reflect the optimum load impedance to the diode.

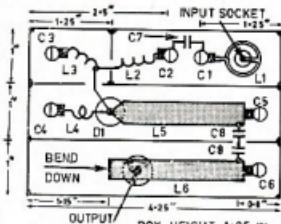


FIG.3. UNDERNEATH LAYOUT-QUADRUPLER

Note: No dimensions are critical, however all joints must be soldered along their full length.

Construction.—The multiplier is constructed in a box of 22 s.w.g. brass, the dimensions of which are given in Fig. 3. The box is first made in the shape of a U and then partitions, coils, tuned lines and finally end plates are soldered on.

Careful consideration must be given to the type of trimmers used. Several types have been evaluated, but the only ones found satisfactory were 6 pF. glass

and 8 pF. ceramic tubular types available through a U.S. disposals source.[†] At 40 watts input, locally available types either caught fire, seized, or shattered.

Alignment.—Connect a 2 metre transmitter of output lower than the rated dissipation of the diode used. It is often unsatisfactory to tune all adjustments for maximum output into a power meter. A better approach is to tune for maximum output at 576 MHz, using a 576 MHz receiver or a cavity filter

¹ John Meshina, Jnr., P.O. Box 62, E. Lynn,
Mass. 01934 U.S.A.

and power meter. Best results are achieved using a spectrum analyser.

Performance.—When correctly tuned, the multiplier produced 9w. at 25% efficiency using a 2N3632 transistor collector-base junction as the varactor. Using an MA4060A, 10w. was obtained at 33% efficiency. Diodes similar to the MA4060A are available for US\$5.00 from a disposals source.¹

AN IMPROVED 576 MHz. VARACTOR MULTIPLIER

The circuit already described suffers from the disadvantages of difficulty of tuning and poor efficiency. Both of these problems can be overcome by the use of a doubler-doubler arrangement, using two diodes. The circuit is shown in Fig. 4.

This system takes advantage of the increased efficiency of the doubler sections. Each doubler operates at an efficiency of about 70%, giving an overall efficiency of 50%. A further advantage of this design is potentially higher power handling capability, however this could not be realised in the author's multiplier due to voltage breakdown of the piston trimmers above 40 watts input.

Further advantages are simple peak adjustment of all variable capacitors and lower spurious output. On-air tests with 10w. a.m. input revealed no detectable distortion. With 40 watts



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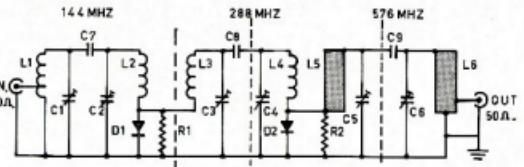


FIG. 4. 144-575 MHz DOUBLER-DOUBLER.

L1=8 turns 18 s.w.g., $\frac{1}{2}$ -inch i.d., tapped 1 $\frac{1}{2}$ turns from cold end, spaced $\frac{3}{8}$ -inch.
 L2=6 turns 18 s.w.g., $\frac{1}{2}$ -inch i.d., spaced $\frac{3}{8}$ -inch.
 L3=3 turns 18 s.w.g., 5/16-inch i.d., spaced $\frac{3}{8}$ -inch.
 L4=2 $\frac{1}{2}$ turns 18 s.w.g., 5/16-inch i.d., spaced $\frac{3}{8}$ -inch.
 L5=2 $\frac{1}{2}$ turns 18 s.w.g., 5/16-inch i.d., spaced $\frac{3}{8}$ -inch above box.
 L6=2 $\frac{1}{2}$ turns 18 s.w.g., 5/16-inch i.d., spaced $\frac{3}{8}$ -inch above box.

f.m./c.w. input, 20 watts output was obtained at 576 MHz.

Physical layout of the improved design is given in Fig. 5 and can also be seen from the photographs. Basic dimensions are the same as for the single-diode design.

CONCLUSION

The designs presented provide ready means of generating more c.w. power on 576 MHz. than can be conveniently generated with valves, and with considerably less complexity.

C1 to C6=1.6 pF, glass piston or ceramic trimmers (see text).
 C7=3.3 pF, ceramic.
 C8=2.2 pF, ceramic.
 C9=0.5 pF, ceramic.
 D1, D2—MA4060A, BAY96, or similar.
 R1=22K Vaw.
 R2=56K Vaw.

REFERENCES

- (1) Motorola Application Note AN147, "High-Power Varactor Diodes: Theory and Application."
- (2) Motorola Application Note AN191, "Varactor Diodes and Circuits for High Power Output and Linear Response."
- (3) Turner, R. P., "ABCs of Varactors" (Foulsham-Sams).

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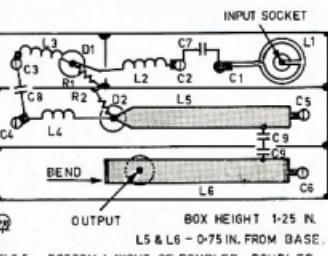
70th ANNIVERSARY OF OLD "CC" TO BE OBSERVED BY W1SS

The year 1971 marks the 70th anniversary of the start of construction of the old "CC" original Marconi station on Cape Cod, Massachusetts. This was the first wireless messages between England and the United States were exchanged by President Teddy Roosevelt and King Edward VII of England.

Those stations desiring to work the site of the original Marconi station will find W1SS active on all bands from 160 metres through 10 metres during the DX hours for each band on the last Saturday of each month. W1SS is the Club Station of the Bedford Massachusetts Radio Club on 24th and 25th April, 1971.

Following is a list of the frequencies W1SS will use:

Band	C.W.	Phone
160 Mx	1.801 MHz.	1.805 MHz.
80 ..	3.580 ..	3.925 ..
49 ..	7.100 ..	7.260 ..
20 ..	14.050 ..	14.155 ..
15 ..	21.100 ..	21.375 ..
10 ..	28.100 ..	28.700 ..
6 ..	50.200
2 ..	145.100



F10.5 BOTTOM LAYOUT OF DOUBLER-DOUBLER

WINTER V.H.F. AND U.H.F. CONTEST

Editor "A.R." Dear Sir,

In order to foster an interest in winter time v.h.f. and u.h.f. operating, I am running a Contest for Australian Amateurs on the bands from 52 MHz. and above.

The duration of the Contest is from 0001 hours E.A.S.T. 1st July, 1971, to 2359 hours, 31st July, 1971.

RULES

1. There is only one division—Transmitting.
2. All Australian Amateurs may enter for the contest whether their stations are fixed, portable or mobile.
3. All Amateur v.h.f. and u.h.f. bands may be used, but cross-band contacts are prohibited. Cross mode contacts will be permitted.
4. Only one contact per band per station is allowed each E.A.S.T. calendar day. Should two or more licensed Amateurs operate any particular station, each will be considered a separate contestant and must submit a separate log under his own call.
5. Entrants must operate within the terms of their licences.

6. Cyphers. Before points may be claimed for a contact, serial numbers must be exchanged. The serial numbers of the first or second figure will be made up of HS (telephone or RST (c.w.)) report plus three figures, commencing at 001 for the first contact and increasing by one for each successive contact.

7. Ineligible Contacts: (a) On the 52 MHz. band, contacts using the mode usually referred to as "Sporadic E" will be disallowed. The sponsor reserves the right to make decisions in doubtful cases.

(b) Contacts over distances below 50 miles on the bands 52 to 580 MHz. will be disallowed as well as those over 25 miles on bands 1215 MHz. and above.

(c) Contacts on net frequencies or through repeaters will be disallowed.

8. Scoring: for all contests will be based on multipliers, multiplied by a factor dependent on the band being used, as follows:

Band	Factor
50 MHz.	1
144 ..	2
432 ..	3
576 ..	4
1215 .. and above	6

Each log entry must show the claimed mileage and score. In the event of two stations disagreeing on the mileage, the average of the two entries will be taken.

9. Logs: All logs must contain the following information: Date and Time (E.A.S.T.), Band, Emission and Power, Call Sign, RST/No. Sent, RST/No. Received, Distance, Points Claimed.

10. Trophy will be awarded to the winner, and consolation prizes will be awarded if the number of entries is sufficient or if any contest results in an Australian record being broken.

ADDITIONAL NOTES

Contestants will observe that the scoring table is wholly based on mileage, including metres. This has been made possible by the disqualification of "Sporadic E" contacts which only occur infrequently at this time of the year. It was felt that this type of contact does not fit in with the operation of "state of the art" equipment and that it was fair to those Amateurs working with meter scatter techniques to also allow "Sporadic E" contacts.

The multipliers are based on the capabilities of Australian stations using "state of the art" equipment or techniques and are roughly inversely proportional to the distances which can currently be expected at that time of the year on each band.

The minimum distances are based on the normal operating range of beginner type stations running 10 watts output to relatively small (by today's standards) antennas, except on 1215 MHz. where 2 watts output is considered more realistic.

References:

- (1) D. W. Bray, K2LMG, "A Method for Determining V.H.F. Station Capabilities," "QST," Nov. 1961, pp. 36-41.
- (2) W. Smith, WIDVE, "Closed Band DX on 52 Mc.," "QST," May 1967, pp. 74-78.
- (3) E. Jamieson, VK3LP, "Meteor Scatter Operations," "A.R." Oct. 1970, p. 24.

Entries to the above Contest should be sent to:

D. D. TANNER,
LIVE & DIXON ROAD,
RIPPLEBROOK, VIC. 3818.

to be posted not later than 31 August, 1971.

Yours faithfully,
D. D. Tanner, VK8AU.

Table 1.—Some available varactor diodes and transistors which can be used as varactors.

* By measurement.

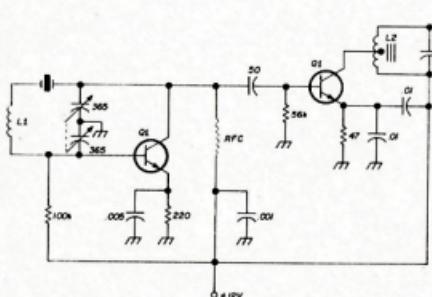
PRACTICAL VXO DESIGN*

An Interesting Approach to Frequency Stability in Oscillator Circuits

GUS GERKE, K6BIJ

You're on the air having an enjoyable conversation. You switch over to the other station and the fellow says, "Sorry, missed most of that. Someone drifted onto your frequency." Sound familiar? The "someone" is usually a combination of unstable v.f.o.'s and receiver drift.

The drifting signals one hears today suggest that v.f.o. stability is not really as good as claimed by equipment manufacturers and authors of v.f.o. articles in the Amateur magazines. The best answer I've found to this problem is the variable-frequency crystal oscillator, or vxo.



The vxo circuits described in this article combine the flexibility (within limits) of a v.f.o. with the inherent stability of crystal frequency control. Frequency can be varied between 2 to 720 kHz., depending on the crystal frequency and other considerations, which I'll discuss. Many Amateurs I have talked to never heard of varying a crystal's frequency over such a wide range.

Very little information has been written about the vxo. One article describes a circuit that can pull down the frequency of an 8 MHz. crystal about 4.5 kHz. before the circuit becomes "a rather inferior v.f.o.". With this circuit (Fig. 1) as a starting point, I designed the circuits of Figs. 2 and 3, using FT-241 crystals in the 450 kHz. region and the circuit of Fig. 4 using 3.5 - 8.5 MHz. crystals.

CIRCUIT DEVELOPMENT

The vxo shown in Fig. 2 is a modification I made to a BC604 f.m. tank transmitter. The vxo output goes through a stage of amplification and several frequency multipliers to obtain output on 21 MHz. I have used this vxo on 7 and 21 MHz. c.w. with excellent results. The circuit has also been used to operate a 2 metre transmitter. Eight crystals were needed to cover the entire 2 metre band.

The only addition to the BC604 was L1, C1. Capacitor C1 is used to pad the crystal frequency over a certain range, in this case 2 kHz. With an increase in padding range, the effects of temperature, vibration, and hand capacitance become more pronounced; and the same precautions in building v.f.o.'s must be used. These effects are small, however, and the crystal is still the frequency-controlling element. If you don't exceed the padding range, the vxo won't become an "inferior v.f.o."

The circuit of Fig. 3 seems to work well with the same low-frequency crystals used in the vxo of Fig. 2. The

Fig. 1.
Circuit described in Reference 1. An extension of 4.5 kHz.¹⁸ claimed for an 8 MHz. crystal.

L1—16-24 uH, for 8-9 MHz. crystal.

L2—40 turns of No. 36 gauge wire, tapped at 16 turns.

Q1—2N706, 2N2219, 2N3662 or R.C.A. 40237.

Table 1 gives recommended padding ranges for the FT-241 crystals when used in the circuits of Figs. 1 through 3. If you are interested in a particular frequency range (as for net operation), try to use a crystal that will cover the first 25 per cent. of the padding range — then you'll have crystal stability.

The transistor circuits will start oscillating with 2.4V.; for more output, up to 12V. can be used. Unless followed by a frequency-multiplier, a buffer amplifier will be needed, as in Fig. 1.

FT-241 Crystal (MHz.)	Fig. 1 (MHz.)	Figs. 2 and 3 (MHz.)
0.45 (fundamental)	0.20	2.00
4.00 (9th harmonic)	2.00	20.00
8.00 (18th harmonic)	4.00	40.00
144.00 (324th harmonic)	72.00	720.00

Table 1—Padding ranges.

A VXO FOR EXCITER USE

Suppose you want to design a vxo covering the entire 40 metre band and you have an exciter such as the Central Electronics 20A using a 9 MHz. crystal.

Higher than 9 MHz. injection frequency is preferred to avoid unwanted mixer products. Therefore the injection frequency will be from $7 + 9 = 16$ MHz. to $7.3 + 9 = 16.3$ MHz. Crystals in this range are overtone types and won't operate in these circuits. The solution is to use an 8.150 MHz. crystal and operate it on its second harmonic, 16.3 MHz. Padding 50 kHz. on the crystal fundamental frequency will produce 100 kHz. shift in the output. This will give you full coverage of the 7 MHz. phone band. An 8.1 MHz. crystal will cover the next 100 kHz., and another crystal at 8.075 MHz. will extend coverage to 7 MHz.

Crystals with frequencies of 8.125 and 8.075 MHz. will be useful if you want extra stability and don't wish to pad more than 50 kHz. on harmonics.

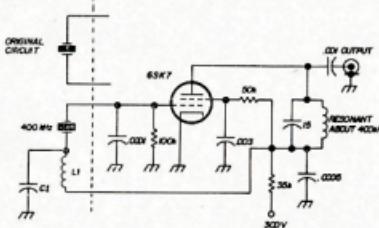
solid state version shown was also used with the BC604. Since the crystals furnished with the BC604 are less than 2 kHz. apart, continuous coverage to the next lower-frequency crystal is possible. Stable 2 kHz. padding was obtained with the circuit of Fig. 3.

A transistor vxo that produces stable 50 kHz. padding is shown in Fig. 4. This vxo can also be used with a crystal in the 8 MHz. region for 6 or 2 metre operation. Doubling will produce a padding range of 100 kHz. on 14 MHz., 150 kHz. on 21 MHz., with tripling, and 200 kHz. on 28 MHz. with quadrupling. To cover the entire 2 metre band, you'll need eight crystals (500 kHz. padding range).

Fig. 2.—Oscillator modification made to a BC604 transmitter using low-frequency crystal.

C1—Broadcast radio variable with both sections in parallel.

L1—Broadcast variloothstick antenna or similar.



(25 KHz. on the fundamental). These crystals are also useful for 2 metre work.

TUNING CAPACITOR CONSIDERATIONS

Referring to Fig. 5, capacitor C1 is used to bring the crystal frequency within the range of C2. Both capacitors should have a straight-line frequency response as a function of angle of rotation of the rotor plates. This capacitor characteristic is important for vxo calibration and tuning. For example, the tuning capacitors shown in the circuits of Figs. 1 through 4 are common broadcast-band variables. When these are used, frequency decreases slowly at first as the capacitor rotor is turned. Then the frequency change becomes faster, until finally a hairline change in rotor position will produce a 1 KHz. jump. This, of course, is very inconvenient at the lower frequencies. The sketch of Fig. 6 illustrates the geometrical relationship of the stator plates in these two versions of variable capacitors.

In the circuit of Fig. 5, capacitor C2 should be of good quality, otherwise contact-scraping noise will be heard in the receiver; small jumps in frequency may also occur. A capacitor with an insulated rotor is recommended for C2.

CIRCUIT DESCRIPTION

The purpose of R1 in Fig. 5 is to lower the Q of L1. This allows a larger padding range and more stable operation near the low end of the range. If the frequency changes when touching the r.f. choke, the choke is too small. Resistor R2 prevents oscillation at the r.f.-choke resonant frequency.

Use a two-section b.c. variable capacitor to find the exact value of C3 and C4. Then replace the b.c. capacitor with two silver micas. A value of 200 pF. seems right for the circuit.

Battery voltage may be 2.4-12 volts. Higher voltage may result in drift due to heating. I use 6 volts in my vxo.

As far I know, the vxo designs described in this article have never been published before. The circuit for the 20A exciter has been used on 40 and

15 metres in both the c.w. and s.s.b. mode. All reports were crystal quality, and all operators asked for the circuit diagrams; so I've presented them here to share with others. My old v.f.o. has since drifted into the junk box.

REFERENCE

1. J. R. Fisk, WIDTY, "73 Useful Transistor Circuits," '73, March 1967.

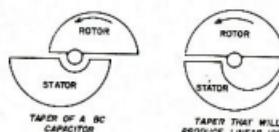


Fig. 6—Mechanical configuration of straight-line wavelength capacitor used for b.c. band and straight-line frequency capacitor.



DX NEWS

It will be noticed that there are no DX notes in this issue. The following letter was received from the DX Editor and it is regretted that NO items of news were received from VK AMATEURS. If this DX page is wanted by readers, more co-operation will be necessary.—Editor.

Editor "A.R.", Dear Sir,

I am afraid we shall have to give the DX page a miss this month. The absence of news from England due to the mail strike has upset the work and only one bulletin has arrived from the U.S.

There has NOT been one item of news from VK this month, and with nill coming in, I guess I can't get anything out.

I hope to have a full page for the following month as I have arranged a fresh source from the States. T3.

—Don Grantley.



EXPEDITION TO LACCADIVE GROUP OF ISLANDS

The Amateur Radio Society of India has sponsored a team headed by Lt. General K. Umrao Singh, VU2US, to visit the Laccadive group of Islands and operate an Amateur station for ten days covering two consecutive week-ends in April 1971. Details are given below:

Operation is expected to start on Saturday, 16th April, 1971, ending on Monday, 19th April, 1971.

Frequencies: 14 MHz. consistently, optional 21 and 28 MHz., both on c.w. and s.s.b.

The rig to be used: 150 watts p.e.p. The call sign will be VUUTUS.

Operators in the party will include VU2CK, VU2QM, VU2HV, VU2KM and VU2RK.

QSL address: Strictly via A.R.S.I., P.O. Box 534, New Delhi-I, India.

Note: All QSL cards will be posted to the L.A.R.U. QSL Bureau, 10, Sector 1, Noida, and string is attached in any shape whatsoever. QSL cards accompanied by IRCs will be mailed accordingly from the A.R.S.I. Enclosing cash currency in envelope is illegal and forbidden according to the country's regulations.

CHANGE OF ADDRESS

W.I.A. members are requested to promptly notify any change of address to their Divisional Secretary—not direct to "Amateur Radio".

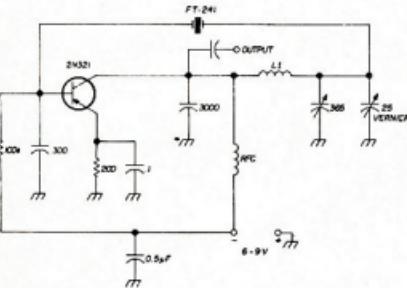


Fig. 3.
Solid state version of the
vxo in Fig. 2.

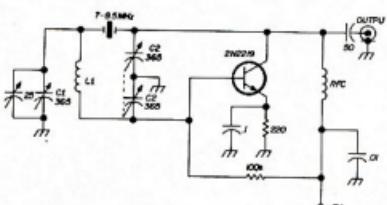


Fig. 4—Solid state vxo that produces stable 50 KHz. padding on 7 MHz. It can be used for 6 or 2 metres also.

L1—40 turns of No. 32 gauge wire close-wound on $\frac{1}{8}$ -inch slug-tuned coil former.

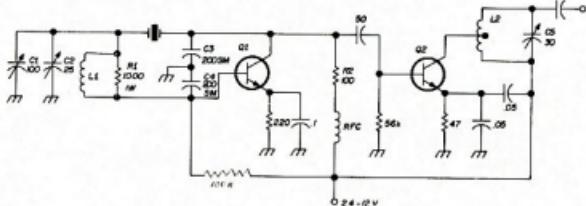


Fig. 5—Vxo-doubler circuit for a typical exciter.

Three crystals are required for full coverage of the U.S. 40 metre band.

Overseas Magazine Review

Compiled by Syd Clark, VK3ASC
and R. L. Gunther, VK7RG

"HAM RADIO MAGAZINE"

November 1970

Editorial—Concerning the new IC, the Signetics N565 monolithic phase-locked loop, a truly remarkable device which can be used for synchronous detection, frequency multiplication or division, I.m. demodulation, and for many other data and applications (from the manufacturer).

Solid State 1360 MHz. Converter, by VK4ZT. Appeared originally in "A.R.".

How to use the Smith Chart, by Jim Fisk.—This is the first time this odd device has made sense to me. Must-reading for any serious Amateur.

Injection Laser Experiments.—Lovely for modulated-light enthusiasts.

Frequency Spotter for General Coverage Receivers.—Simplest transistor crystal oscillator for a 1 MHz. crystal.

Radio Teletype using S.S.B. Transceivers.

Auxiliary Receiver for 100 Metres.—Raises the maximum frequency of an ordinary b.c. or c.w. receiver.

A Counter Gating Source.—Using the mains frequency as a gate for frequency counters.

Voltage Regulation using the R.C.A. CA-3885.

Very Versatile and Impressive.

Linear V.H.F. Tank Circuits.—The design of simple single ended v.h.f. tanks using quarter wave transmission lines on 2 metres.

Printed Circuit Boards without Printing.

The old trick of removing copper from a board by use of an electrician's drill (Pardon the chloroform) is much easier. If you find it difficult to apply "resist", simply use black felt-tip pen ("Texta-color" or equivalent), wash off with acetone after etching. But his idea is good; it is better to leave copper in insulation, leaving most of it on the board for low inductance and shielding.

A Simple Test Set for Transistors and Diodes.

Uses a couple of diodes and lamps. Not very good, but it does accomplish as much by judicious use of an ohmmeter (but not on its lowest ohm scale).

December 1970

As usual the decimal points in the diagrams are often vanishingly small, and can cause real (and dangerous) confusion. One hopes that this problem will be solved by the Editor at an early opportunity.

Filter-Type S.S.B. Generator, WSKIT. Another one, transistorised. The balanced modulator and filter certainly make a simple circuit to do. The accompanying statement is impeccable: "Why not try a little construction? There's nothing to compare with the satisfaction gained in creating something worthwhile."

Noise, Radio Frequency Interference, WIDTY. The usual noise sources and their cures (at the source), with a special word about fluorescent lights: chokes as big as 2.5 MHz. are needed. In general, a good job is done, particularly the information provided by the mains, at least in Australia.

The R.F. Bridge, WB2EGZ. Very good. Same as the "Antenna Noise Bridge" described in "H.R." in 2/69, and the author has had no trouble putting to good use his E Noil (WB3PQJ) in his interesting series on antenna construction, published by Sams. People still persist in using simple v.s.w.r. bridges even though they solve little and can mislead greatly. The balanced bridge, on the other hand, is not much more complicated, and is well described in this issue of "H.R."

Avalanche Transistor Circuits, WINVK. Quite ordinary transistors can be made to show alternating frequency response and power when operated in the avalanche mode (though this will not work with some of the older types having low avalanche voltages). Example, a 5 mW. unit can produce a 1000 p.p.m. 1-2 MHz. rise time. To produce an avalanche, merely apply some 200 V. to a transistor in common-emitter mode through a large resistor and drive the base over the conduction threshold at 100 MHz. The output signal is tuned to WWV; it will determine the accuracy of all of the harmonics to some 1000 MHz., obtained from the avalanche unit. A 5 KHz. drive gives 59 signals to 30 MHz.!! Avalanche can be generated at low voltages simply by

turning the transistor upside down. Most interesting.

Low-Power Transmitter and Indicating Wave-meter, WENIF. A keyed one-transistor oscillator plus diode detector with amp.

A Symmetrical Frequency Synthesizer Oscillator for RTTY, WFOO. The generation of pure tones to encode s.s.b. transceivers for r.t.t.y.

Identifying Unknown Transistors, W2FPP.

The usual ohmmeter tests, but don't use the ohmmeter to measure the forward current as it passes substantial current. Amplification and Beta are tested by substitution into an R/C oscillator or an amplifier, but in my opinion it's easier to use the ordinary bias-free method (e.g. A.R. 11/68, p. 69). The most useful suggestion he gives is to test r.f. performance in a crystal oscillator; see also "Break-In," 3/68.

Harmonics, Distortion and Splatter, K5LLI.

See the better series being run currently in "A.R." (Or at least recently, if it has terminated when this appears.)

Improved Super-regenerative Receiver, by JA1BHG. Not surprisingly, transistor action in the super-regen is used to supply impedance matching, e.g. tapping collector and emitter on the tank coil. A diode across the emitter tap reduces hangover (see "H.R." 11/68), but in my opinion you still need an r.f. amp. to reduce the distortion. An s.s.b. is all very broad. Scope for application of the simple super-regen is, alas, becoming ever more restricted on the fecund amateur spectra.

A Flexible Voltage-regulated Power Supply, WSEK. Uses an IC with ten connections. The current-limiting is on the wrong side!

Should be on the supply side, not in the feed-back loop on the load side! This article is needlessly uninformative.

The Ham Notebook (Letters): Resistors can be frequency sensitive above 10 MHz. or so; also the author's note in late 1970 issue of "Spectrometer" (N.Z.) Beware.

A WWV (or Lyndhurst) converter can be made to cover Amateur bands by heating WWV against a suitable crystal.

Yackar (and other) oscillators will oscillate better on low d.c. voltages if the load resistor is replaced by a choke (but beware of spurious resonances).

A pi-net can be used to match indoor antennas to commercial rigs; what won't they think next?

An fine deviation-indicator can be made simply by tuning one receiver 15 KHz. (etc.) away from another, and feeding both into a discriminator feeding a c.r.o.: ingenious.

An argument whether an inverted ve has a higher or lower resonance frequency compared to a dipole. It appears to depend on the height of the ends of the vee above the ground, the electrical length decreasing as the antenna becomes lower.

Another interesting (as well as an authentic) has triumphantly discovered that shorting out a portion of a tuner coil will absorb r.f. power. One wonders whether people bother to learn basic circuit theory nowadays before obtaining an Amateur licence; the phenomenon is hardly restricted to the Yankee nation.

On top of all this, the D.J. Dorn, DW3JJ, has published ample material in "Ham Radio," "73" and "E.E.B." to show that on one hand a coupler is virtually necessary for the best possible in-band performance reduction. How the use of a coupler could be harder than cutting lengths of line, is a mystery; that it is more "expensive" is a statement argument for partially obvious reasons.

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has triumphantly discovered that shorting out a portion of a tuner coil will absorb r.f. power. One wonders whether people bother to learn basic circuit theory nowadays before obtaining an Amateur licence; the phenomenon is hardly restricted to the Yankee nation.

Eimac, and presumably the other manufacturers too, will be applying the IVS rating to material designed for Amateur use, but there is no reason why we couldn't also establish our own values for IVS rating on other valves used in Amateur practice, such as t.v. bins, output meters, etc.

Modifying the Heath SB-200 Amplifier for the New 843 Zero-Bias Triode, W6UDV (also of Eimac). This is one of the valves for which Eimac has prepared IVS ratings, and these were discussed in the above-mentioned article.

Tow Metre F.M. Frequency Meter, WA1AJZ. A highly accurate heterodyne frequency meter giving crystal-controlled frequency markers on 2 metres.

100 W. Power Amplifier for 200 MHz. WB-6DJV. Uses a 7854 valve with very simple construction and high performance. Possible substitute is the 5634, but it could require modification, which it appears the 7854 does not, if good geometry is employed. D.C. input about 5000 V.

Inexpensive S.W.R. Indicator, WB2GQV. Wind about 9 turns of wire to $\frac{1}{4}$ inch diameter, connect to diode, condenser, and meter in the usual detector circuit. Slip the coil over the transmission line, turn, roll, bend, and forth to determine strengths of loops and nulls. Is somewhat better than the usual in-line v.s.w.r. bridge in that it can also indicate match of the antenna to the line as well as match of the transmitter to the line. But an r.f. noise bridge and explicit antenna coupler are better; see last month's "H.R." And I do disagree strongly with this author who says "it is a more convenient device (than the transmitter) in view of added expense" so that the indicator uses the cut-and-try approach until the transmission line electrical length is close to a half-wavelength or multiple. When this condition is met, the transmitter will lead the antenna to resonance.

When this condition is met, the transmitter will lead the antenna to resonance. It only applies if the antenna is matched by the line, it ought to now be common knowledge that a suitable matching coupler is desirable for a good v.t.-x line matching with minimum difficulty. As far as I am concerned, it is a mystery that a coupler could be harder than cutting lengths of line, as a matter of all-potential for physically obvious reasons.

And on top of all this, the D.J. Dorn, DW3JJ, has published ample material in "Ham Radio," "73" and "E.E.B." to show that on one hand a coupler is virtually necessary for the best possible in-band performance.

For antenna use, it is relatively harmless at ordinary h.f. Isn't it time that Amateurs wake up to the more obvious aspects of v.s.w.r. and coupling?

Five Pictures in the Ham [sic] Shack, Darr and James. Use fuses of the right size; both current and voltage rating. Mains sockets and plug should have clean contacts, particularly for high voltage wiring. Should be adequate. His suggestion is excellent that a special mains line should be run to the radio room, and provide plenty of outlets. This is a vitally important subject, often overlooked.

MOSFET Converter for Receiver Instrumentation, WA2ZMT. Uses a 3N159, feeds a c.r.o. to monitor the l.f. passband of a receiver.

A Simple C.W. Monitor, WA2OH. Two transistors of opposite polarity connected in the regenerative feedback mode, powered from the keyer. Simple indeed.

The Ham Notebook (Letters): An ohmmeter can be used to find the sensitivity of an unknown meter. Only practical with v.t.m. Practical calculations involved, it is just as simple to use a handy dry cell and a few resistors.

A wire coat hanger can be made into a long screwdriver, which need for adjustments in difficult places.

Sensitivity and stability of the 75A-4 receiver can be improved. Mostly by replacing leaky condensers; in any old receiver it is a good idea to replace all condensers on principle.

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RADIO COMMUNICATION

October 1970

The GRARV Two Metre Portable Receiver. An solid-state FET front-end into a tunable 22.5-24.5 MHz. l.f. and thence into a 10.7 MHz. l.f. filter, etc.

Lift Aerials. GM4QK describes means of being an active Amateur without shouting it to the world.

An Antennae Retainer, GCCKV. Some "junk" and lots of ingenuity and now set and forget.

A Simple Transistor Tester, G3NUQ describes a simple test set we could all make and afford.

Technical Topics, G3VA continues his review of the happenings in Radio. "Solid State Receiver Design" by WB0YHH and many more. A warning is sounded against the constant use of disc ceramic capacitors as "decouplers".

without checking that they are doing the desired job. Seems some of them resonate as low as 23 MHz.

November 1976

Current Comment concerns itself with the need of the R.S.G.B. to increase subs to £4 (A\$8.60). Inflation is no respecter of countries.

Parasitic Oscillations in V.h.f. Power Amplifiers. G. S. M. Teale. Reprinted from "Mullard Technical Publications". Solid state circuits are under discussion.

An R.F. Indicator for the Blind. G7TA. An aid for those without sight.

Compact 150W. Amplifier for 144 MHz. G6JGP. A 4CX320 in a grounded cathode circuit.

A Simple 3 CM. Polplexer. G3EEZ. The klystron is a 2SK2 or 723A.

The G3XGX Vacker Oscillator. Various circuit configurations are discussed. The author claims there is little to choose between them.

A Portable W. Transceiver for 3.5 MHz. G2ZL. 2N708, 2N709, 2N706, 2N705, 2N704, 2N719 diode ring modulator, BC169, BC169, BC169, 2N708 is the receiver line-up.

Technical Topics. G3VA. Pat Hawker discusses the latest information to appear in the various journals which are available to him. His "T" is much lengthier than this review and is a valuable source of the technical articles Pat considers worthy of his attention. This month is devoted to Direct Conversion, d.c.-d.c. converters and co-axial reed relays amongst others.

December 1976

A 1-16-100 KHz. Calibrator. G3UCM. The article enlightens the reader somewhat more than the title for what the author really means by this. The calibrator provides signals at intervals of 1, 10, and 100 KHz. to 30 MHz. at least.

Obtaining Deviations. G3EDD. F.m. or p.m. valve or transistor.

Flare Spot. Part 1, Crime Wave. G3BGL presents a radio-detective story in three parts.

Technical Topics. G3VA. K2QBW multi-band aerial-based verticals, lof loop serial, aerials, etc. What do they differ, some static superhet, ideas, two-wire notch filter, r.f. how do you turn a toroid? (see p.835) r.f. power transistors and broadband amplifiers, audio filters, simple linear time base.

Modifications to the HW-100, SB-100 and the SB101.

"SHORTWAVE MAGAZINE"

October 1976

Getting on VFO for the VHF/UHF. G2JF. A valve type v.f.o./driver system beginning with a 2-tube 2N4420, 2N4421, followed by the Hartley Clapp oscillator circuit. A system of heterodyning the v.f.o. second harmonic on about 7 MHz. with a signal of 132 MHz. from a multiplier chain is used to provide output on 145 MHz.

Reed Relay Reflection. G3TMG. A critique of the article by VK1AU which was published in "S.W. Mag." August 1976.

Varactor Diode Circuits. G3TNX. Theoretical considerations and some practical circuitry.

VFO TX for Twenty. G3HVV. Describes an experimental QRP rig which can be duplicated easily. Costs: Five 2N2222 transistors into a pi network which cost the author 30/- (in U.K.) and gave 160 QSOs in 42 countries over a few weeks.

Design of Linear Amplifiers. G6HL. A discussion of the use of 4CX250 and 4CX350 types for h.f. and v.h.f. work.

"73 MAGAZINE"

November 1976

Differential JFET Pre-amplifier.—A highly commendable endeavour to use discrete components to achieve superior results by borrowing from IC technology.

Delta Tuning.—An ingenious arrangement is employed to allow reflector stub tuning directly at the operator's position, by bringing the reflector termination down to a tuning capacitor via a feed line.

With the Help of the Repeater using the Heterodyne V.F.O.—Instead of the crystal-heterodyne v.f.o. (which appeared in a previous "73") you can feed any source of 120 mW. 50 MHz. to drive this simple but effective common-emitter, C.R. amplifier with v.w. output.

Semiautomatic E.M. Channel Selector.—A repeater frequency is switched automatically between two channels, by use of a "flasher" module, a simple multivibrator which drives transistors at the oscillator cathodes.

Low Cost Amateur Keyer.—Another one, using only five transistors.

A.C. Switching with Self-Powered ICs.—R.C.A.'s CA3049 allows turning a Triac on and off by switching only when the a.c. voltage passes through zero.

Radio Hobbies on the Prairie.—Rather interesting tale on the activities of E. E. Krebsbach, 45 years ago.

The SST-1: Solid State Transceiver for 40 Metres.—The receiver follows the May 1969 "WCE" design, a direct conversion design using an R.C.A. IC.

A Low-Cost E.F. Wattmeter.—The usual diode probe r.f. voltmeter.

Calibrate Your V.H.F. Calibrator.—Beat the calibrator until V.H.F. is correct, then adjust "crystal tune" until the S meter stops passing. Allows sub-sonic zeroing, with b.f.o. off.

Study Guide; General Class Licence, Part IV.

A \$10 Noise Generator.—Back bias a u.f.b. diode across a 100 ohm resistor, through a large resistor. Zener diodes also work at lower freq.

(Note)—This magazine commonly contains a large amount of editorialising by various parties and numerous controversial letters. Although the publication is exoteric, the cogent points are raised, and if you have a spare hour or two, it can make interesting reading.)

December 1976

Solid State Exciter, W8YUY. S.s.b. for the home-brewer with plenty of test equipment.

Delta Solid State Control of S.S.B. Exciter, W4NVK. A varcap vernier frequency control.

A 2 Metre Minimitter for Repeaters, WB6BBH. One watt output, 18 MHz. xtal. 22.5 volt supply.

Receiver Offset Tuning for HW-100, WAB2AEW. A link coupled remote tuning system for v.f.o.'s.

The Little Gate Dipper, W8ETT. Another simple 1.7 to 225 MHz. g.d.o. with MPF102.

Your Second Linear, W8AYL. Uses 3-56Z.

General Class Study Guide, Part 5. Valves.

Yipes It Talks, W2FEZ. Make your own electrostatic loudspeaker out of a newspaper.

Transistor Test, W8BQP. A very simple BJT leakage shorts tester.

Two-Terminal Current Limiter, Gerald Beene. A two-legged fuse to protect the three-legged variety.

January, 1977

The usual editorial tirades, and some interesting notes on how Japanese industry is undercutting the manufacturers of commercial amateur equipment (N.I.P.). And a number of interesting letters from readers.

LX for Leisure, G3BID. If you would like a leisurely vacation with a bit of amusing radio operation thrown in, you might consider Luxembourg.

Try Xing the World . . . the Hard Way. K6KA. A magnificent world tour costing "thirteen thousand kilobucks" in which the new and unusual sights included numerous antennas and operating positions.

Split Phases, A DX Operating Aid, GW8GP. GW8GP discovers dual diversity with headphones. Good idea.

A Special Report, Ham Radio Manufacturing: A Struggle for Survival, W2NSD. The question is: will Amateur Radio survive? "The ham-equipment manufacturers could meet foreign makers head-on in their own territory . . . V.H.F. I.m. is proving to be very popular, needed shot in the arm for home radio sales."

Heath Tener Modification, KA1JKL. A major engineering project; the correct size fuse is obtained automatically when the Heath Tener is installed in the automobile.

Duty Cycle Duty Factor, W2OLU. Good introduction to the concept of duty cycle (percentage of time the key is held down). But unsatisfying.

The real implications for the choice of valves or transistors are left unsaid; see "Intermediate Voice Operation of Power Transistors", W8SAL, in "73" January 1976, for the real deal.

Repeater Zero Beater, W1RHR. Monitors the repeater receiver's discriminator voltage during a transmission and stores a voltage in a condenser. When the transmission is completed the voltage is converted to a tone which is transmitted during the repeater's fall period. Interesting idea.

Getting HEP to ICs, Staff. Tips on wiring soldering, cross references, and simple projects using the Motorola HEP Integrated Circuits.

Voices from the Past, Staff. Quotes from Amateur Radio 50 years ago, Radio 30 years ago, and "73 Amateur Radio" 10 years ago. Those good old days?

Basis of Surius F.M., WB2AEB. How to buy your own.

A Parabolic Beam for 10, 15, 20 Metres, by W2EZZ. Very good. Works on the same principle as the standard reflector, but it's non-dimensional. Uses aluminum tubing and No.

12 wire as support for aluminium sheet metal strips. Works on all three bands, with forward gain from 14 to 25 dB, depending, and 40 dB. F/B is driven element in any good quality vertical monopole. The antenna is, of course, the author recommends a commercial unit. But he does not mention the effect of wind in a good blow. I'd use bonded glass fibre.

The Galaxy FM210, K2ULR. A 2 m. f.m. transceiver described to be as good as a Japanese one but have been changed, but not ever since. And, my word, the FM210 is all-American, parts and labor. Unless there have a magician in the engineering section, something will have been compromised in this wondrous feat. But it does look quite attractive.

Lightning as it Affects Ham Radio, Patsch. Very good. Install a suitable safe lightning arrester on the mains and also run stranded down wires (4 gauge!) from the antenna tower on your house, down to a good earth connection. See "Lightning Protection and Fire Protection" in Jan. 1971 "Ham Radio."

IC Receiver Accessory, W3EZY. Plugs into a headphone jack, runs a loudspeaker, a.g.c., and tunable a.f. filter. Slight adjustment of the a.g.c. feedback resistors may be necessary, but subsequent readjustments will quickly settle this inside the IC. A do-it-yourself project.

Inverted Attic Antennas, W2SF. Inverted vee in an attic, uses electrical conduct for elements. Okay with lots of American output power, but house wiring can absorb energy when efficiency matters.

Double Balanced Mixer, K3PUR. A survey of different types. The transistor double balanced mixer looks interesting and requires no balanced transformers.

Quick and Permanent Tool Marker, K5JKX. 24k, through the tip of a draughtsman's pencil. But it's easier if you connect the lead at the other end of the pencil—a long piece doesn't have to be lost at a time. Also works if you use any piece of sharp metal as an etched scribe, but control is a bit harder.

A New Start from Washington, W8GL. A bitter attack on A.R.R.L. and its Establishment. But it's truly grateful that the lead at the end of the W.I.A. is tractable as they are.

Amateur Radio License Study Guide, Staff. With such a comprehensive list of requisite technical subjects as are presented here, it does seem rather a pity that so many of the licensees are destined to apply their knowledge to assembling commercial plugs, IC and antennas.

FEEDBACK

I am indebted to Ron VK5OM for pointing out that my remark in the review published in "A.R.R.L. Jan. 1971" in relation to an article in the Oct. 1970 issue, "Designing an External VFO for the SB10 'Transceiver'" was incorrect. I was thinking of the Sideband Electronics Raytheon solid state transceivers SB-33 and SB-34. These transceivers were designed to cover quite a small segment of the Amateur bands as "standard"—VK3ASC.



HY-Q ELECTRONICS EXPAND

Hy-Q Electronics Pty. Ltd., Australia's leading quartz crystal manufacturers, have announced that, as a result of ever increasing demands for their products from Australia and overseas, major expansion of their production bands has been necessary.

Hy-Q have recently completed the construction of a modern, fully air conditioned plant located at 1 Rosella St, Frankston, Vic., devoted entirely to the production of quartz crystals and related frequency control products which has tripled the company's previous production capacity.

The new plant has been equipped throughout with the most modern crystal production and testing apparatus including equipment for fine scale testing, cold well crystals.

The new facility includes a separate, fully equipped emergency service without disrupting normal production.

The company's original factory at 19-22 Rosella St, being converted to provide fully air conditioned development laboratories, engineering shops and office facilities.

W.I.A. V.H.F.C.C.

New Member: Conferences

Cert. No.	Call	22 MHz.	144 MHz.
78	VK4ZJB	120	—

New Equipment

YAESU FT-101 SOLID STATE TRANSCEIVER

Some time has elapsed since the Yaesu Musen Co. Ltd. of Japan produced their first solid state transceiver, model FT-100. The present model, FT-101, basically similar, incorporates the latest advances featuring 10 FETs, 3 integrated circuits, plug-in modules, noise blanker, as well as 31 silicon transistors and 38 silicon diodes. The transmitting section employs 3 tubes only, a 12BY7A driver and 2 x 6JS6A final amplifier with an output on s.s.b. of approx. 160 w.p.e.p.

The built-in dual power supply provides for operation from alternative power sources, 12v. d.c. or 234v. a.c. Selection of the appropriate power cord, from the two provided, is the only adjustment for a change-over.

A desirable feature in a set such as this is the built-in speaker. A matching external speaker, external v.f.o., c.w. filter and mobile mounting hardware are available as optional extras. It covers the usual Amateur bands of 80-10 metres, plus the 11 metre band, and includes reception of WWV on 10 MHz. Modes of operation are s.s.b., c.w. and a.m. C.W. input power is adjustable. Panel meter indicates p.a. cathode current, r.f. output, and a.l.c. On receive, the meter functions to read "S" units.

Taking into account the advantage of low current drain, the FT-101 is the perfect choice for use in a car, caravan, boat, aircraft, and field day activity. It also excels as a primary base station.

Of special interest to brass pounders, c.w. operation is a real pleasure with near perfect keying characteristics, absence of chirp, stability, high selectivity, and "break-in" with side tone monitoring.

A photo appears elsewhere in this issue, and full details are available from the Australian agent, Bail Electronics Services of 60 Shannon St., Box Hill North, Vic. 3129.

GEELONG "HAMFEST" OVER THE WEEK-END OF 1st and 2nd MAY, 1971

Saturday: 1400 hours onward, registration and rag-chew. Dinner and entertainment.

Sunday: Displays of commercial gear, scrambles and tx hunts on 40 and 2 metres, barbecue lunch, disposals sale, entertainment for everyone.

Further details from VK3 W.I.A. Broadcast or the Geelong Amateur Radio-T.V. Club Secretary, Bob Wooley, VK3IC, P.O. Box 520, Geelong, Vic. 3220. Telephone 212674.

Book Review

SINGLE SIDEBAND FOR THE RADIO AMATEUR

Over the last twenty years the A.R.R.L. has done a great deal to popularise s.s.b. amongst the Amateur fraternity; nowadays one hears more s.s.b. than c.w. or a.m. signals, especially on the DX bands and some amateurs have been led to believe that the s.s.b. operators will not talk to them.

We live in a rapidly changing world, exciting things are happening somewhere in the world every day of the week and the rate at which science is advancing is said to double itself every ten years.

The fifth edition of Single Sideband for the Radio Amateur will assist the newcomer to our hobby in becoming acquainted with the mode and bring the old-timer up to date on the more modern techniques. Sixty per cent of the material is new and heavy emphasis has been accorded to solid state devices.

This issue contains thirty-one practical constructional projects from easy-to-build station accessories through simple receivers to the more sophisticated crystal filter and phasing type exciters, transmitters and complete transceivers.

This new edition contains 250 pages and is \$12.95 post paid. Price is \$3.30 post paid from the W.I.A. Federal Executive Publications Department or Divisional Secretaries.—VK3LC.



FROM THE W.I.A. NOVICE INVESTIGATION COMMITTEE

The following extracts are taken from a letter on the subject of Novice Licensing, received from Mr. William L. Orr, WSA1, a prominent technical writer in the field of Electronics and Amateur Radio. His opinion may be of interest to Novice Licensees. He should offer some valid arguments to those who are in favour of such a licence on the Australian scene.

"Generally speaking, the Novice programme has been a healthy one in the U.S.A. No general opposition exists to it. In most New Zealand towns (particularly smaller ones) follow the Novice route. It gives them a 'taste' of Amateur Radio and encourages them to carry on. . . . Many of today's prominent Amateurs were Novices at one time, and quite likely many of us were novices ourselves. The Government's examination unless their confidence had been built-up by actual on-the-air contacts and expertise they had gained during their Novice period.

"The Novice concept was introduced by the Federal Communications Commission over the objection of the A.R.R.L. The A.R.R.L.'s personal opinion was that the A.R.R.L. was afraid that this 'sub-standard' licence would degrade Amateur Radio. Fortunately, this did not happen, and I am positive today that the A.R.R.L. supports and encourages this programme.

"Change is always difficult and hard to accept, especially in organisations which tend to reduce all to the lowest common denominator. The Novice licence in the U.S.A. tended to become a ghetto, much QRM, poor operating techniques, etc. Most General class Amateurs avoid the segments, which is a pity. Even so, the Novice learns to tune a transmitter and receiver and a bit of experience. Some of them do quite well."

"The great danger to Amateur Radio is not the Novice class, but the unfortunate monster created by Citizens Radio—2,000,000 licensees and many, perhaps, operators. This has caused Amateur Radio on growth, as many would-be Amateurs take the easy road to communicate by radio via the C.B. route, rather than by the more demanding road to Amateur Radio. For anyone who has real interest in Amateur Radio, I hope anyone who has real interest in Amateur Radio, should be encouraged in every way possible."

"So many interests are available to the young—t.v., motor bikes, autos, travel, marjorians, and thousands of other interests in electronics should jump for joy when a youngster evidences interest in Amateur Radio. Tomorrow's communicators and electronic engineers are coming from this pool of manpower. It is important that we, as Amateurs, do our bit to keep interesting and constructive hobby and advancement. I know of no better way of doing this than to appeal to the timid newcomer by means of a Novice class licence—the first rung on the ladder of Amateur Radio."

—R. C. Black, VK2YYA, Chairman.
[We publish the foregoing as a matter of interest. We do not necessarily agree with all Mr. Orr's observations.—Ed.]

FEDERAL REPEATER SECRETARIAT

This month we are pleased to be able to include a report from the Gold Coast Radio Club on the first fully operational Channel 1 system in Australia. We invite the technical officers of other repeater groups to submit a report along similar lines about their own system, both for our own records and publication in "A.R."

The first report for 1971 from the F.R.S. has been produced and has been sent out. If we have missed any groups and you would like a copy, write to the F.R.S. c/o, P.O. Box 342, Crewe Nest, N.S.W. 2065.

CHANNEL ONE SYSTEM ON QUEENSLAND GOLD COAST

The Gold Coast Radio Club, as a club project, has established an FM repeater station to service the South East Queensland and North Eastern N.S.W. areas. The repeater has been April M.G. licensed and fully operational since April 1970. Details of the repeater are as follows:

Call Sign: VK4EI/R2.
Frequency: Repeater Channel 1 (146.1 MHz. in and 145.9 MHz. out).

Location: Mt. Tamborine. Approx. 18 miles west of Southport and 40 miles south-west of Brisbane. Site elevation is approx. 3000 ft. a.s.l.

Tx: Complete valve design, multiplying from 4 MHz. xtal. 25 watts carrier output from QQE606 4-p.t. valve. The power output is soon to be boosted to 50 watts when construction of a new tx is complete.

Rx: Solid state throughout, realising 20 dB signal to noise ratio for 0.5 microvolt p.d. input with the tx carrier on. 1st i.f. is 10.7 MHz. and second is a Pyle 40-70 kHz filter. 2nd i.f. is 455 KHz.

Aerials: Both tx and rx use identical aerial types comprising five half-wave collinear elements, fed in phase, vertically polarised, omnidirectional and realising 8 dB power gain. Both aerial systems are mounted 40 ft. above ground level and are horizontally separated by 250 ft. The aerials are fed via a pair of cavity resonators in both the tx and rx feed lines has reduced rx desensitisation to almost nil.

Availability: The repeater is available on a 24-hour basis. The rx runs continuously and when the repeater is opened the tx and rx are keyed on. Eight minutes after the squelch opens closed, the tx files are keyed off. Following the initial squelch opening, each successive squelch operation returns the tx files shut-down time delay to zero.

Identification: Automatic station identification after a five-minute "carrier-on" duration. Solid state keys for morse code identification are presently being experimented with.

Coverage: Good service is available within a 200 mile mile circle centre on the repeater site. Good mobile to mobile QSOs have been conducted between the following areas: Lindsay, Byron Bay, Brunswick Heads, Tweed Heads, Brisbane, Gold Coast, North Coast resorts, Murwillumbah, Boonah and many other places.

Well that is roughly the story regarding the Gold Coast repeater. A repeater for Brisbane is still being considered by the VK4 V.H.F. Group, but as yet no sign of air testing. Channel 4 will be used for Brisbane's repeater and will be known as "RI" until an official call sign is allocated.

The Gold Coast Radio Club will be only too happy to pass on information regarding the project to inform other groups of the pitfalls and other aspects involved in a repeater. A note to D. J. Adams, VK4ZDA, Gold Coast Radio Club, P.O. Box 388, Southport, Qld. 4215, will ensure full technical details, etc., by return mail.

Recently a copy of Ken Sessions, Jr. K6MVH's "Radio Amateur's F.M. Repeater Handbook" arrived in Australia. This is an excellent publication on the subject, but much of the material applies only to the American scene, where many Amateurs have full control of their stations—usually on a suitable hill-top. Other chapters are devoted to the "tone" control and access to repeaters, a system which does not apply in this country. Tim VK2ZTM, Chairman F.R.S.

AMATEUR FREQUENCIES:

ONLY THE STRONG GO ON—SO
SHOULD A LOT MORE AMATEURS!

VHF

Sub-Editor: ERIC JAMIESON, VK5LSP
Forreston, South Australia, 5233.
Closing date for copy 30th of month.
All Times in E.S.T.

AMATEUR BAND BEACONS

VK3	53.544	VK3GR	Antarctica.
VK3	144.700	VK3VE	Vermont.
VK4	144.390	VK4VV	107m. W. of Brisbane.
VK5	53.000	VK5VF	Mt. Lofti.
VK6	144.490	VK6VY	Warrnambool.
VK6	52.096	VK6VF	Tuart Hill.
	52.900	VK6TS	Carnarvon.
	144.500	VK6VE	Mt. Barker.
VK7	144.900	VK7EV	Tuart Hill.
VK9	144.600	VK9VF	(by arrangement).
ZL3	145.000	ZL3JVF	Christchurch.
JA1	51.905	JA1IGY	Japan.
W	50.081	WB6KAP	U.S.A.
HL	50.100	HL5WI	South Korea.

Only change to the beacon list this month is the corrected location of the VK3 beacon to Vermont, Mt. Koothal. I was previously informed, March and April will be months to keep a look out on 6 metres for long distance contacts, particularly to JA and other northern areas so the beacons listed here in those areas may be the frequencies to monitor in the shade over a week-end and doing some constructional work!

Six metres appears to have been comparatively quiet during the past month, a few signals from VK5AU and VK5ZWW. VKA advises he and Doug VK5KKK in Darwin now have their equipment operating well enough on 32 MHz. VK5 to allow themselves easily being able to send one another early morning skeds. Copy is usually about 10 s.s.b. with occasional 5 x 2 periods which allow a reasonable exchange of information. Just around sunrise seems to be the best time. Doug has now started running Saturday and Sunday mornings starting at VK5ZWW from 0630 to 0700, calling each alternative five minutes. David starting his transmission first. Signals have been quite good some of the time and successful exchanges were made three out of the last four attempts.

David worked JA1MRS on 2nd Feb. and on 24th Feb. JA5, HL5WI and HK5CR. Similar situation of signals being heard from same areas on 17th Feb. (it seems some DX might be lost in the contest—VK5LSP).

I note David is interested in promoting v.h.f. winter activity and it appears he has sent details for publication in "A.R." of a winter v.h.f.-u.h.f. contest, probably during the month of July. I certainly hope it will receive a full venture, but will leave further comment until I know more about the proposed contest.

Thank you for your letter, David, always pleased to hear of activities in the north. Long time since I heard from Doug VK5KKK, or anyone in North Queensland!

Well, the John Meyle National Field Day Contest has been and gone. A number of stations in VK5 went out portable, but extended range contacts, particularly 2 metres, were very few and far between. My own effort was confined to 2 metres this year, for a few hours on Sunday morning. However, I was pleased to be able to include the boys managing the portable station on Mt. Arapiles (the Orange Radio Society) by working VK5BES, VK5ZYT and VK5YBMM, and seem to have been the only VK5 to have done so from this area. I was rather staggered to find no stations available from the south-east of VK5, except down the coast the VK5ZWW, Bob 2ARH, NM5NN, Roy SAXY, Roy SAOS, Eric Z2KN, Jim JAEEF and so I could go on. It did not seem like a v.h.f. field day without these calls. I know conditions were not real good, looks like the call of the h.f. bands is too strong!

Fortunately, temperatures in VK5 were much lower than the Adelaide peak of 105 last year, but strong south-westerly winds tended to slow some of the lower bands. Looking eastward from the south, it appears that best contact to be made during the N.F.D. was probably that between Norm VK5ZQC portable on Mt. Tassie near Taralgon and Eddie VK1VVF portable on Mt. Gingera, Victoria. A distance of about 100 km. over a rather mountainous path. Good work chaps, shows the effort was worthwhile for a 2 metre contact.

My faithful sidekick from VK3, Bob VK3AOT, sent some further useful information this month. He advises that John VK5AJM and

David VK3ANP (?)—the pen's not too sharp Bob!—are fully operational on 432 MHz. and are set up to take part in the Australian Open S.B. contest on 11th March. Another and another from the same area currently constructing 432 MHz. gear is Peter VK5JPP.

1326 MHz. RECORD

Bob VK3AOT passes on the news that the 1326 MHz. record has been broken again, on 17th Feb., when Ron VK3AKC in Geelong and Kevin VK7ZAII in Launceston established contact over a path of 274.3 miles, bettering the previous record by more than 20 miles. Both stations were on 1326 MHz. contact with similar reports around 55. The equipment at VK7ZAII comprises a C239 tripler producing about half a watt output and a 5 feet by 2 feet parabolic section antenna. After the initial contact, the test signal from VK3AKC was copied for 90 minutes by VK7ZAII and VK3JVV. Congratulations go to both for a great effort, and particularly to Ron, who previously only held the record for six hours! A late report indicates that the stations were in contact on 1226 at 1755 on 27th Feb. VK7ZAII being received at 54 and VK3AKC at 86. Ron will soon be running out of suitable territory in Tasmania if he continues to push his signals south. Let us all hear what he has moved to Lakes Entrance and concentrating signals on the path to New Zealand!

Bob continues his writing with some excerpts from the latest release from the P.M.G. Dept. showing the growth rate of amateur radio services for the 1968-70 financial year. Amateur full licences increased by 6.1% and limited licences by 8.3%. The overall growth rate of radio communication stations was 16.1%, while that portion of the service in the same period was 2.0%. Some sobering thoughts come out of those figures if you care to reflect for a moment. Many thanks again Bob for your contribution.

A further reminder to the Geelong Amateur Radio and The Club Hamfest scheduled for 1st and 2nd May. No further news of this event has arrived but no doubt separate information will be made available through "A.R."

From the West Australian V.H.F. Group, Noel Birkin comments that the VK5GWF went portable for the N.P.D. at the Eagle Hill Forestry Lookout. Activity was definitely multiband, c.w. on 3.5 and 7 MHz., s.s.b. on 15, 14, 12, 10 and 28 MHz., a.m. on 7, 17, 20 and 24 MHz., f.m. on 10 and 14 MHz. An article on the subject of the beacon, noted also in the same publication, advice to the effect that the 2 metre beacon has been recently modified and now at 5 watts has more trouble than its previous power. The 10 metre beacon has also been strengthened up somewhat since its old 6/40, which was half-dead, has been buried!

HL5WI WORKED IN VK5

News has just trickled through for this Stop Press item that HL5WI in South Korea was worked over a wide area on 1st March. It appears a total of five VK5s were worked, the only one mentioned by name so far being Peter VK5ZWW. While contacts were also made to Doug VK5ZWW, VK5ZYT and VK5YBMM. The band was still open to the north at 1326 and there is an unconfirmed report of a VS8 working into VK5. David VK5EAU reported hearing the HL5WI beacon on 52 at 1000 which is generally the last contact or hearing of this nature. So it looks as though March and April could be interesting months for 6 metres, as mentioned earlier.

A brief note to hand mentioned Wally VK5ZWII and VK5ZYT were successful in having a 30-minute contact via meteor scatter on 28th Feb. So it looks as though the Saturday and Sunday morning skeds are paying off. Bob VK5ZDX is now operational on 52 MHz. c.w. and is finding a transverter with a QQ2E6/40 in the final. This is designed to run into a high powered linear and when finished he hopes too to join in the M/S experiments. Wally VK5ZWII, by the way, has been planning to go to France and prepared to run some skeds with him using M/S. He is rather interested in finding someone in the Eastern States with s.s.b. gear preferably, who will turn the beam in his direction.

No news has come to hand about any portable operation during Easter, so various areas will need to rely on their weekly broadcasts for such information. But I do suggest if you are in the shade over Easter keep a sharp eye on 6 metres with the beam north, particularly during the afternoon and early evening periods—you may well hear something worthwhile, as the Easter period will probably bring more visitors to the air than usual. If you are seeking something to monitor, try JA1IGY on 59.750 MHz.

News has been somewhat scarce this month, only two letters received, but there may be more next time. In closing here is the thought

for the month: "A church is a hospital for sinners, not a museum for saints." Until next month, 73, Eric VK5LP. The Voice in the Hills.

MEET THE OTHER MAN

Meet Wally Watkins, VK5ZWW, of Bellevue Heights, a suburb of Adelaide on the slopes of the Mt. Lofty Ranges, at an elevation of 750 feet, living amongst the elite and able to hold court on most of the population of Adelaide.

Wally formerly was ZL2TCW, living at Lower Hutt, New Zealand, and several years ago came to Australia with his wife Dorothy and family. There seems to be no evidence of any of them arriving in chain! Wally was first licensed in 1963 and while in New Zealand was a keen DX enthusiast, particularly on 144 MHz., from where he worked ZL1 and 3. While VK5 and VK6, the latter under pseudonym, he had contact with Huile VK5KS a distance of 1950 miles, running 30 watts a.m. From his present location on 52 MHz. he has worked VK5, VK6, 4.5, 5, 6, 7 and 8, JA1, 2, 3, 4 and 5. On 44 MHz. areas worked are VK3, 5, 6 and 7 and contacts have been made within VK5 on 432 MHz.

Equipment in use at present on 52 MHz. uses a QQ2E6/40 in the final, running 150 watts d.c. and 100 to 90 elements up 30 feet. He uses a VK5 FET converter. The system changes to a.m. for 144 MHz. and runs 190 watts to a 52B9, modulated by a pair of 2N174s, and coupled to a 5/8 slot antenna up 30 feet. A 500-watt a.m. transmitter uses a 52C4 front end. Back to a. 1. for 432 MHz., running 15 watts to a 5939, 16 element collinear up 30 feet, VK5 FET converter. The tunable if, in each case, is a FWD100.

Wally is a member of the W.I.L.D. (Western Institute of Amateur Radio), Committee for 1970, and supplies the v.h.f. notes for the VK5 Journal. He operates portable from time to time on 52 MHz. when his work as a surveyor with the Commonwealth Government takes him to suitable areas and occasionally tries 144 MHz. At present is actively interested in 52 MHz. scatter, and has successfully worked David VK5EAU in Tennant Creek on a number of occasions using this method. His future plans include attempting to work all stations on 52 MHz. meteor scatter and with this in mind is planning to increase power on 52 MHz. to the legal limit. He also is planning equipment for 144 MHz. n.t.a.

VKA finds Wally a worthwhile asset for Amateur Radio, one who likes to see things done, and is not afraid to speak his mind at meetings, treading on a few corns in the process no doubt, but his words are meant well. Pleased to have you living with us, Wally.



John VK5ZQJ adjusts a 10 el. 2 metre yagi. Top antenna is a 15 el. 432 MHz. top. Beneath John are a pair of 146 MHz. vertical 10 el. yagis and a 4 el. 6 metre yagi. The tent in the background housed a 2 kva motor-generator set. These aerials were in use by VK5AOT/P at Mt. Cowley from 18/12/70 to 2/1/71.

COOK BI-CENTENARY AWARD

The following additional stations have qualified for the Award:

Cert. No.	Call	Cert. No.	Call	Cert. No.	Call
1149	DJ62T	1167	EABBN	1226	VE3YEJO
1150	AX5NM	1184	K4CIA	1227	FBIQ
1151	AX7PS	1189	AZ2DA	1228	HL9WT
1152	AX1ACZ	1190	AZ2KM	1229	VE3LA
1153	AX1EIR	1191	AZ2LW	1230	AX3HE
1154	EASGK	1192	JAI1KA	1231	WSUMR
1155	TM2BK	1193	AX3AJX	1232	WB4LFV
1156	PA9QT	1194	JAI1DVN	1233	W4ELB
1157	AX2VN	1195	DK2D	1234	VE3X
1158	TM2G	1196	DK2SWY	1235	JAS1F
1159	ZS6VC	1197	JAS1KZ	1236	AX2ARV
1160	ZM1QC	1198	ZM1HV	1237	9G1FF
1161	AX2AW	1199	W2FCR	1238	FR2TZG
1162	JAI1DXK	1200	W2KQ	1239	W4P
1163	AX3HM	1201	W2XZ	1240	SK3EC
1164	WS5KD	1202	AX2ABC	1241	AX3HK
1165	DL3EM	1203	AX3AMU	1242	AX3XJ
1166	IIACY	1204	AX2ANZ	1243	VE3FFM
1167	JAI1D	1205	DK2D	1244	VE3LE
1168	WV2GF	1206	DL2BK	1245	W3VGH
1169	AX1AIG	1207	WB2NYM	1246	WUJMW
1170	W3AZD	1208	W6AAK	1247	W8AJW
1171	W4ATP	1209	MPMBH	1248	UAGDG
1172	AX3EDC	1210	W2KQ	1249	W4P
1173	GS2ED	1211	WB2JYM	1250	W6DCK
1174	W8AG	1212	VE1AH	1251	SM0CER
1175	WT1AW	1213	W4EWR	1252	YU3ZV
1176	WA9HOM	1214	ZM4MW	1253	HL9WT
1177	WV2DQE	1215	ZM2DQ	1254	W4P
1178	JAI1OIE	1216	LG1XZ	1255	AX3IAKS
1179	VSC8CW	1217	OZ4EZ	1256	ZS1MH
1180	DJ5JY	1218	JR1BEK	1257	WU2CWK
1181	OE3RHA	1219	JH1EX	1258	EA1ANI
1182	KG1ZGZ	1220	JH1EX	1259	VE3SNW
1183	DL3ECL/	1221	G3RUX	1260	DK1XA
W2		1222	AX2AVZ	1261	YU2NEG
1184	YV3UF	1223	VE3CN	1262	CR3OE
1185	WB6BKD	1224	W5XZO	1263	JA2NDQ
1186	GMSNW	1225	FSLK	1264	HS3ADW

V.H.F./U.H.F. SECTION

Cert. No. Call
20 AX3AOT
21 AX7ZRO

W.I.A. D.X.C.C.

Listed below are the highest twelve members in each section. Position in the list is determined by the first number shown. The first number represents the participants who counted less any awarded credits for deleted countries. The second number shown represents the total D.X.C.C. credits given, including deleted countries. Where totals are the same, listings will be alphabetical by call sign.

Credits for new members and those whose totals have been amended are also shown.

PHONE

VKMMS	319/343	VK4PJ	287/297
VKRU	317/343	VK4TY	284/288
VK3AHO	311/326	VK2APK	281/287
VK3AII	311/326	VK2AII	281/287
VKAKS	362/317	VK3TLL	271/277
VK5AB	297/314	VK3ZE	265/268

New Members:

Cert. No.	Call	Total
115	VK1IDK	231/231
116	VK3AII	102/102

Amendments:

VK2SG	258/260	VK4RF	198/199
VK4UC	253/353	VK2AIIH	208/208
VK3VK	238/326	VK3TGF	175/179

C.W.

VK3QL	303/326	VK3NC	274/300
VK3AIIQ	361/315	VK3XK	270/287
VK3AIIH	361/315	VK2AIIH	270/287
VK3AGH	252/298	VK6SU	268/269
VK3APK	250/288	VK4TY	259/272
VK3YL	280/287	VK3TLL	255/268

Amendments:

VK4KF	175/187	VK2AIIH	135/144
VK3SG	141/145		

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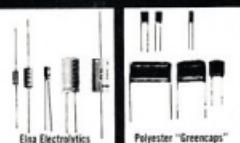
VK5RU	318/343	VK4KS	303/322
VK2AGH	314/321	VK2R	302/325
VK2EVN	300/328	VK3AIIK	299/308
VK3AII	300/328	VK3AII	299/308
VK4TY	306/321	VK4RJ	298/323
VK5MK	304/324	VK2SG	294/300

Amendments:

VK4UC	282/283	VK2AIIH	214/223
VK4RF	240/252		

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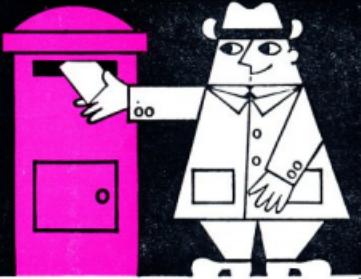
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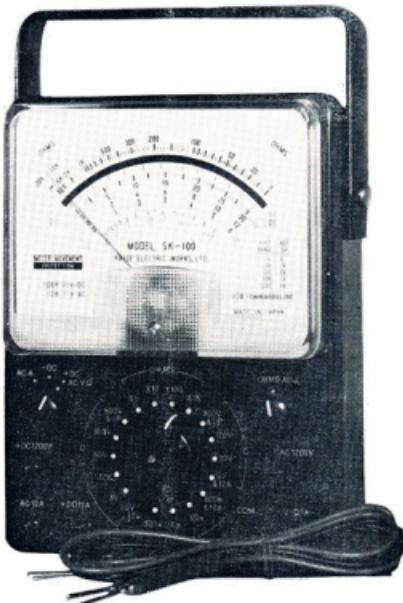
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